



INTEGRATED TECHNICAL EDUCATION CLUSTER  
AT ALAMEERIA

**J-601-1448**  
**Electronic Principles**

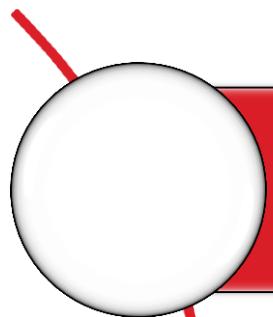
# Lecture #11

## Feedback Circuits

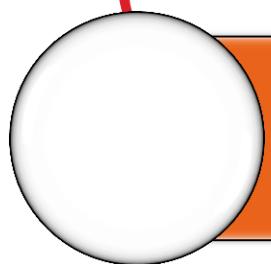
**Instructor:**  
**Dr. Ahmad El-Banna**



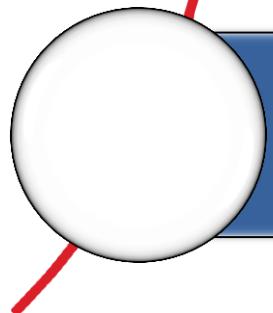
# Agenda



Feedback Concepts



Feedback Connection Types

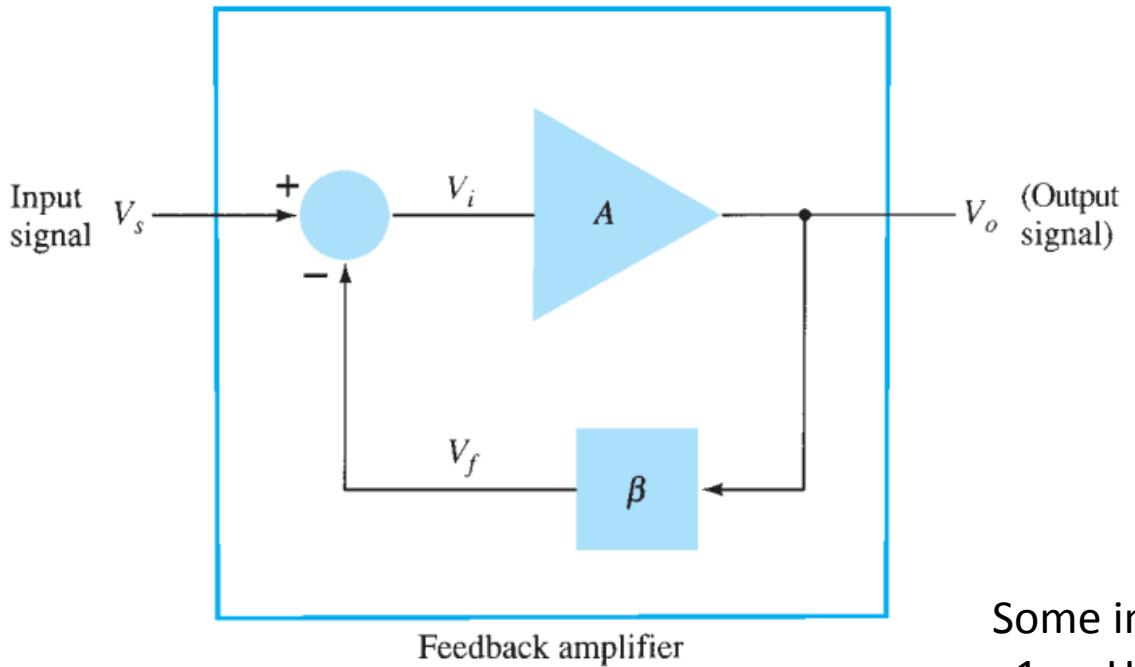


Practical Feedback Circuits

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# Simple block diagram of feedback amplifier

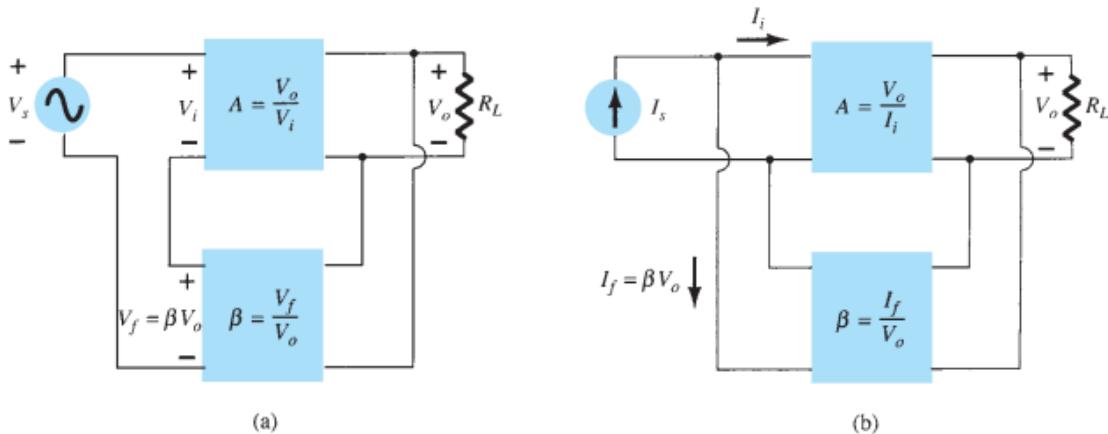


Some improvements are :

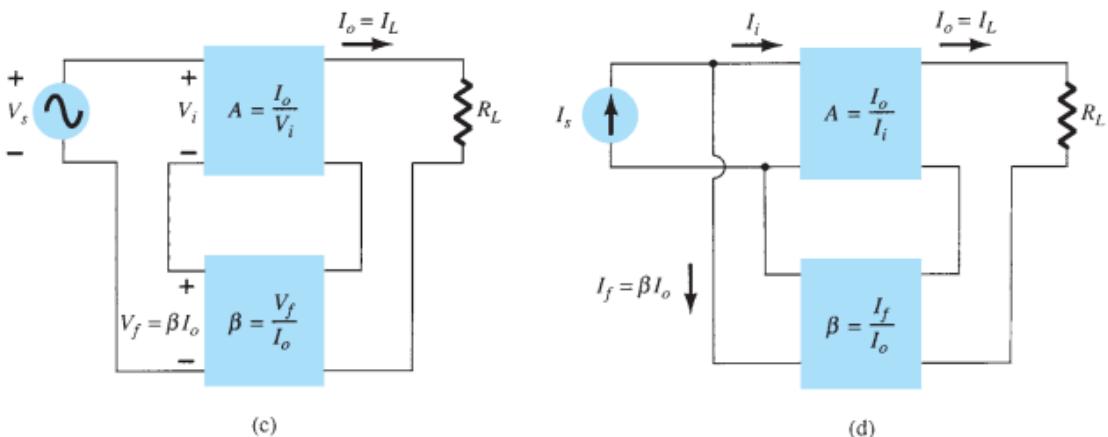
1. Higher input impedance.
2. Better stabilized voltage gain.
3. Improved frequency response.
4. Lower output impedance.
5. Reduced noise.
6. More linear operation.

## FEEDBACK CONNECTION TYPES

1. Voltage-series feedback (Fig. 14.2a).
  2. Voltage-shunt feedback (Fig. 14.2b).
  3. Current-series feedback (Fig. 14.2c).
  4. Current-shunt feedback (Fig. 14.2d).



- Series feedback connections tend to increase the input resistance, whereas shunt feed-back connections tend to decrease the input resistance.
  - Voltage feedback tends to decrease the output impedance, whereas current feedback tends to increase the output impedance.



**FIG. 14.2**

*Feedback amplifier types:* (a) voltage-series feedback,  $A_f = V_o/V_s$ ; (b) voltage-shunt feedback,  $A_f = V_o/I_s$ ;  
 (c) current-series feedback,  $A_f = I_o/V_s$ ; (d) current-shunt feedback,  $A_f = I_o/I_s$ .

# Gain with Feedback

**TABLE 14.1**

*Summary of Gain, Feedback, and Gain with Feedback from Fig. 14.2*

		Voltage-Series	Voltage-Shunt	Current-Series	Current-Shunt
Gain without feedback	$A$	$\frac{V_o}{V_i}$	$\frac{V_o}{I_i}$	$\frac{I_o}{V_i}$	$\frac{I_o}{I_i}$
Feedback	$\beta$	$\frac{V_f}{V_o}$	$\frac{I_f}{V_o}$	$\frac{V_f}{I_o}$	$\frac{I_f}{I_o}$
Gain with feedback	$A_f$	$\frac{V_o}{V_s}$	$\frac{V_o}{I_s}$	$\frac{I_o}{V_s}$	$\frac{I_o}{I_s}$

## Voltage-Series Feedback

$$A = \frac{V_o}{V_s} = \frac{V_o}{V_i}$$

$$V_i = V_s - V_f$$

$$V_o = AV_i = A(V_s - V_f) = AV_s - AV_f = AV_s - A(\beta V_o)$$

$$(1 + \beta A)V_o = AV_s$$

$$A_f = \frac{V_o}{V_s} = \frac{A}{1 + \beta A}$$

## Voltage-Shunt Feedback

$$A_f = \frac{V_o}{I_s} = \frac{A I_i}{I_i + I_f} = \frac{A I_i}{I_i + \beta V_o} = \frac{A I_i}{I_i + \beta A I_i}$$

$$A_f = \frac{A}{1 + \beta A}$$

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# Input Impedance with Feedback

## Voltage-Series Feedback

$$I_i = \frac{V_i}{Z_i} = \frac{V_s - V_f}{Z_i} = \frac{V_s - \beta V_o}{Z_i} = \frac{V_s - \beta A V_i}{Z_i}$$

$$I_i Z_i = V_s - \beta A V_i$$

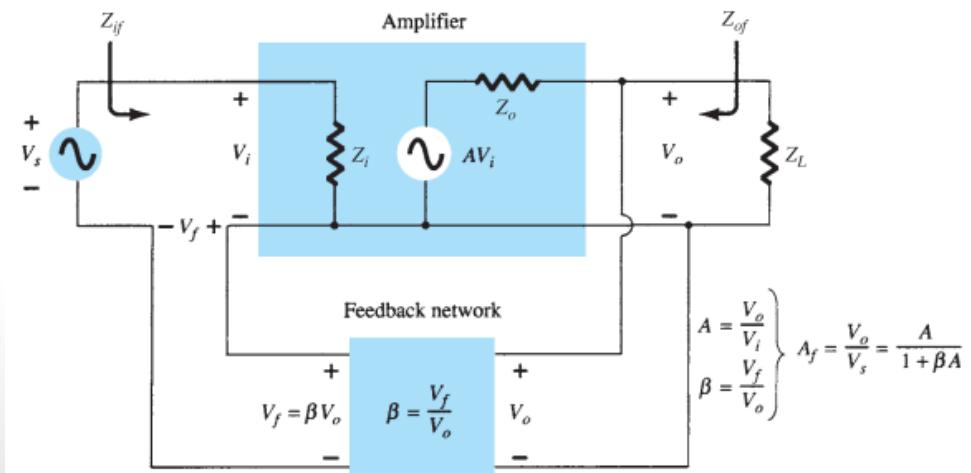
$$V_s = I_i Z_i + \beta A V_i = I_i Z_i + \beta A I_i Z_i$$

$$Z_{if} = \frac{V_s}{I_i} = Z_i + (\beta A) Z_i = Z_i(1 + \beta A)$$

## Voltage-Shunt Feedback

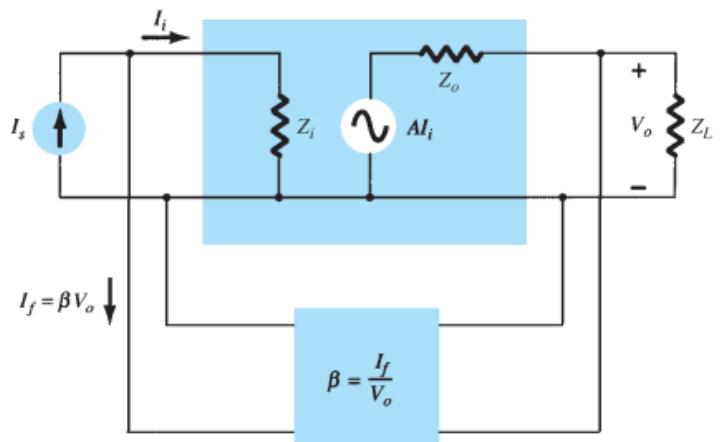
$$\begin{aligned} Z_{if} &= \frac{V_i}{I_s} = \frac{V_i}{I_i + I_f} = \frac{V_i}{I_i + \beta V_o} \\ &= \frac{V_i/I_i}{I_i/I_i + \beta V_o/I_i} \end{aligned}$$

$$Z_{if} = \frac{Z_i}{1 + \beta A}$$



**FIG. 14.3**

Voltage-series feedback connection.



**FIG. 14.4**

Voltage-shunt feedback connection.

# Output Impedance with Feedback

## Voltage-Series Feedback

$$V = IZ_o + AV_i$$

$$V_i = -V_f$$

$$V = IZ_o - AV_f = IZ_o - A(\beta V)$$

$$V + \beta AV = IZ_o$$

$$Z_{of} = \frac{V}{I} = \frac{Z_o}{1 + \beta A}$$

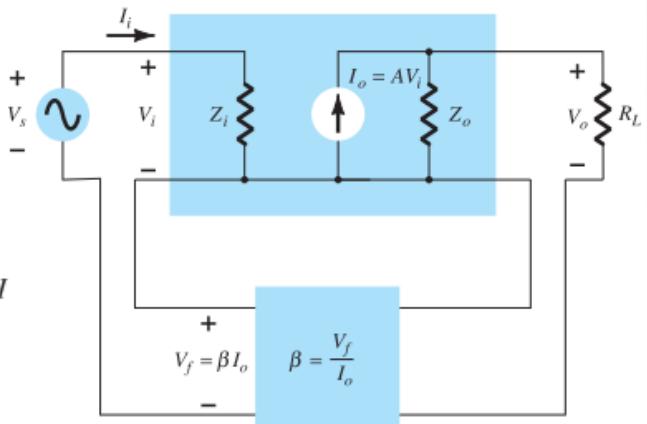
## Current-Series Feedback

$$V_i = V_f$$

$$I = \frac{V}{Z_o} - AV_i = \frac{V}{Z_o} - AV_f = \frac{V}{Z_o} - A\beta I$$

$$Z_o(1 + \beta A)I = V$$

$$Z_{of} = \frac{V}{I} = Z_o(1 + \beta A)$$



**FIG. 14.5**

Current-series feedback connection.

**TABLE 14.2**  
Effect of Feedback Connection on Input and Output Impedance

Voltage-Series	Current-Series	Voltage-Shunt	Current-Shunt
$Z_{if}$ $Z_i(1 + \beta A)$ (increased)	$Z_i(1 + \beta A)$ (increased)	$\frac{Z_i}{1 + \beta A}$ (decreased)	$\frac{Z_i}{1 + \beta A}$ (decreased)
$Z_{of}$ $\frac{Z_o}{1 + \beta A}$ (decreased)	$Z_o(1 + \beta A)$ (increased)	$\frac{Z_o}{1 + \beta A}$ (decreased)	$Z_o(1 + \beta A)$ (increased)

# Example

**EXAMPLE 14.1** Determine the voltage gain, input, and output impedance with feedback for voltage-series feedback having  $A = -100$ ,  $R_i = 10 \text{ k}\Omega$ , and  $R_o = 20 \text{ k}\Omega$  for feedback of (a)  $\beta = -0.1$  and (b)  $\beta = -0.5$ .

**Solution:** Using Eqs. (14.2), (14.4), and (14.6), we obtain

$$\text{a. } A_f = \frac{A}{1 + \beta A} = \frac{-100}{1 + (-0.1)(-100)} = \frac{-100}{11} = -9.09$$

$$Z_{if} = Z_i(1 + \beta A) = 10 \text{ k}\Omega (11) = 110 \text{ k}\Omega$$

$$Z_{of} = \frac{Z_o}{1 + \beta A} = \frac{20 \times 10^3}{11} = 1.82 \text{ k}\Omega$$

$$\text{b. } A_f = \frac{A}{1 + \beta A} = \frac{-100}{1 + (-0.5)(-100)} = \frac{-100}{51} = -1.96$$

$$Z_{if} = Z_i(1 + \beta A) = 10 \text{ k}\Omega (51) = 510 \text{ k}\Omega$$

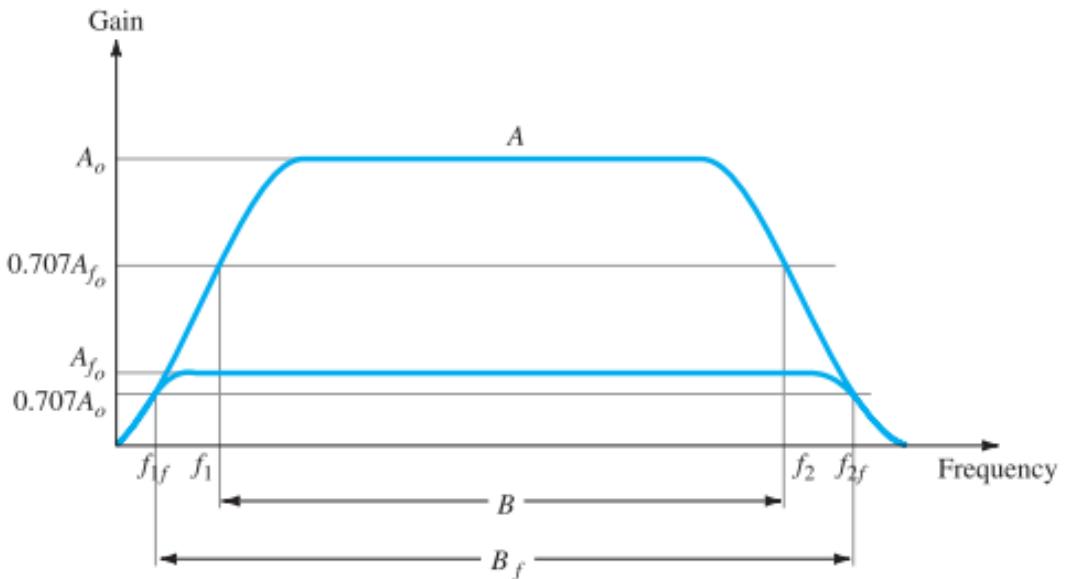
$$Z_{of} = \frac{Z_o}{1 + \beta A} = \frac{20 \times 10^3}{51} = 392.16 \Omega$$

# Effects of feedback on Gain and Bandwidth

## Reduction in Frequency Distortion

## Reduction in Noise and Nonlinear Distortion

$$A_f = \frac{A}{1 + \beta A} \cong \frac{A}{\beta A} = \frac{1}{\beta} \quad \text{for } \beta A \gg 1$$



## Gain Stability with Feedback

$$\left| \frac{dA_f}{A_f} \right| = \frac{1}{|1 + \beta A|} \left| \frac{dA}{A} \right|$$

$$\left| \frac{dA_f}{A_f} \right| \cong \left| \frac{1}{\beta A} \right| \left| \frac{dA}{A} \right| \quad \text{for } \beta A \gg 1$$

# Practical Feedback Circuits

## Voltage-Series Feedback

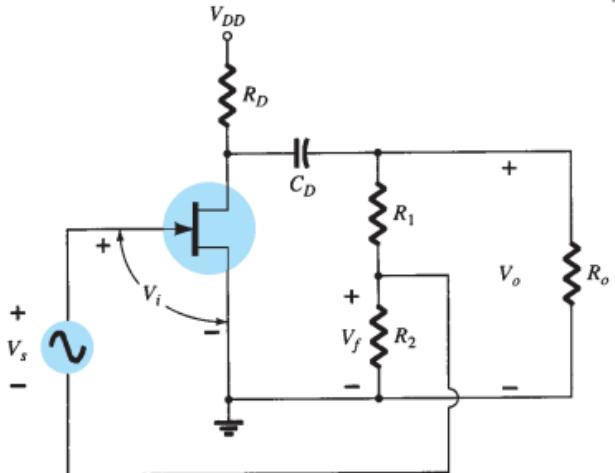
$$A = \frac{V_o}{V_i} = -g_m R_L$$

$$R_L = R_D R_o (R_1 + R_2)$$

$$A_f = \frac{A}{1 + \beta A} = \frac{-g_m R_L}{1 + [R_2 R_L / (R_1 + R_2)] g_m}$$

$$\beta = \frac{V_f}{V_o} = \frac{-R_2}{R_1 + R_2}$$

$$A_f \approx \frac{1}{\beta} = -\frac{R_1 + R_2}{R_2}$$



**FIG. 14.7**

FET amplifier stage with voltage-series feedback.

**EXAMPLE 14.3** Calculate the gain without and with feedback for the FET amplifier circuit of Fig. 14.7 and the following circuit values:  $R_1 = 80 \text{ k}\Omega$ ,  $R_2 = 20 \text{ k}\Omega$ ,  $R_o = 10 \text{ k}\Omega$ ,  $R_D = 10 \text{ k}\Omega$ , and  $g_m = 4000 \mu\text{S}$ .

**Solution:**

$$R_L \approx \frac{R_o R_D}{R_o + R_D} = \frac{10 \text{ k}\Omega (10 \text{ k}\Omega)}{10 \text{ k}\Omega + 10 \text{ k}\Omega} = 5 \text{ k}\Omega$$

Neglecting the 100-kΩ resistance of  $R_1$  and  $R_2$  in series gives

$$A = -g_m R_L = -(4000 \times 10^{-6} \mu\text{S})(5 \text{ k}\Omega) = -20$$

The feedback factor is

$$\beta = \frac{-R_2}{R_1 + R_2} = \frac{-20 \text{ k}\Omega}{80 \text{ k}\Omega + 20 \text{ k}\Omega} = -0.2$$

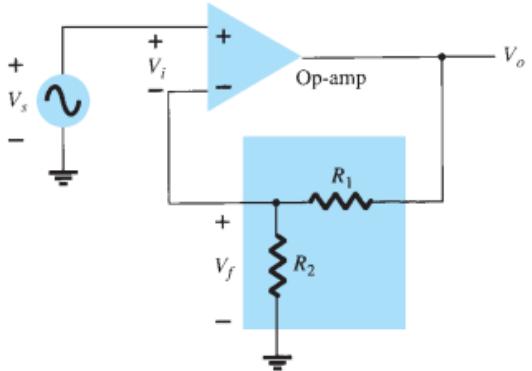
The gain with feedback is

$$A_f = \frac{A}{1 + \beta A} = \frac{-20}{1 + (-0.2)(-20)} = \frac{-20}{5} = -4$$

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# Using OP-Amp & Emitter follower

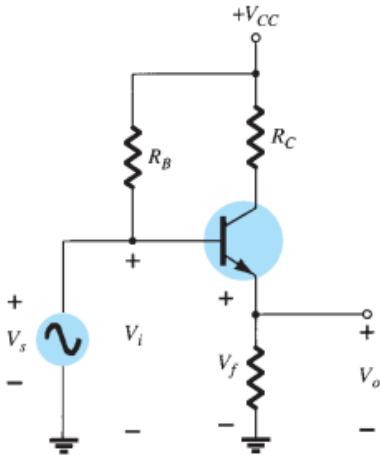
$$\beta = \frac{R_2}{R_1 + R_2}$$



**FIG. 14.8**

Voltage-series feedback in an op-amp connection.

Check EXAMPLE 14.4



**FIG. 14.9**

Voltage-series feedback circuit  
(emitter-follower).

$$A = \frac{V_o}{V_s} = \frac{h_{fe} I_b R_E}{V_s} = \frac{h_{fe} R_E (V_s/h_{ie})}{V_s} = \frac{h_{fe} R_E}{h_{ie}}$$

$$\beta = \frac{V_f}{V_o} = 1$$

$$A_f = \frac{V_o}{V_s} = \frac{A}{1 + \beta A} = \frac{h_{fe} R_E / h_{ie}}{1 + (1)(h_{fe} R_E / h_{ie})}$$

$$= \frac{h_{fe} R_E}{h_{ie} + h_{fe} R_E}$$

For  $h_{fe} R_E \gg h_{ie}$ ,  $A_f \approx 1$

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# Current Series Feedback

## Current-Series Feedback

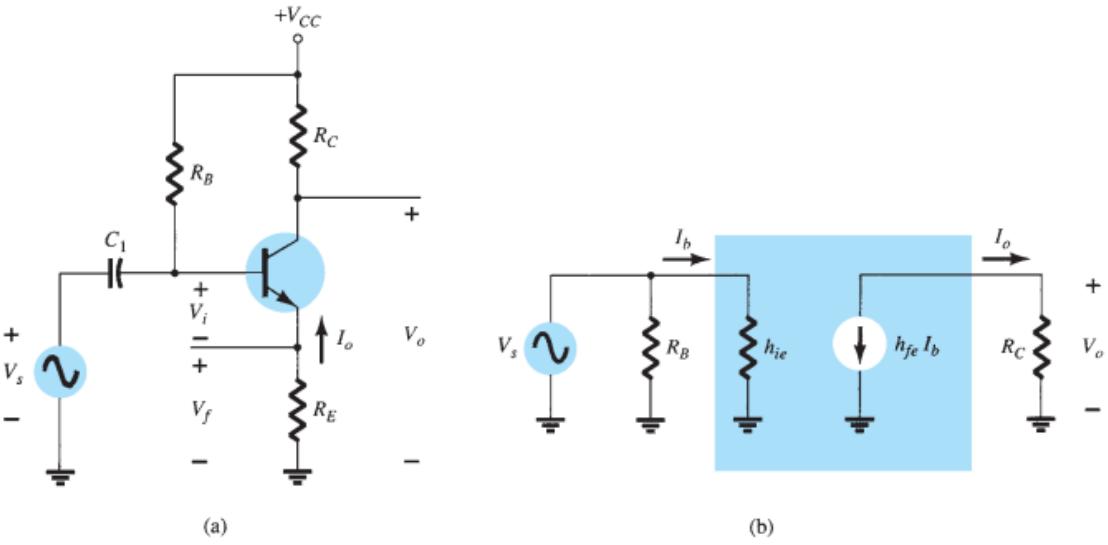
### Without Feedback

$$A = \frac{I_o}{V_i} = \frac{-I_b h_{fe}}{I_b h_{ie} + R_E} = \frac{-h_{fe}}{h_{ie} + R_E}$$

$$\beta = \frac{V_f}{I_o} = \frac{-I_o R_E}{I_o} = -R_E$$

$$Z_i = R_B \parallel (h_{ie} + R_E) \cong h_{ie} + R_E$$

$$Z_o = R_C$$



**FIG. 14.10**  
Transistor amplifier with unbypassed emitter resistor ( $R_E$ ) for current-series feedback: (a) amplifier circuit; (b) ac equivalent circuit without feedback.

### With Feedback

$$A_f = \frac{I_o}{V_s} = \frac{A}{1 + \beta A} = \frac{-h_{fe}/h_{ie}}{1 + (-R_E)\left(\frac{-h_{fe}}{h_{ie} + R_E}\right)} \cong \frac{-h_{fe}}{h_{ie} + h_{fe}R_E}$$

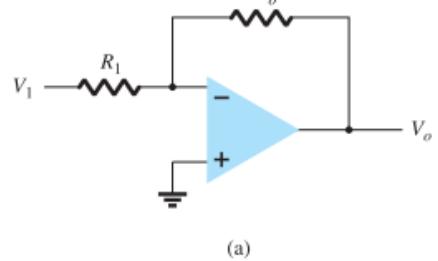
$$Z_{if} = Z_i(1 + \beta A) \cong h_{ie} \left(1 + \frac{h_{fe}R_E}{h_{ie}}\right) = h_{ie} + h_{fe}R_E$$

$$Z_{of} = Z_o(1 + \beta A) = R_C \left(1 + \frac{h_{fe}R_E}{h_{ie}}\right)$$

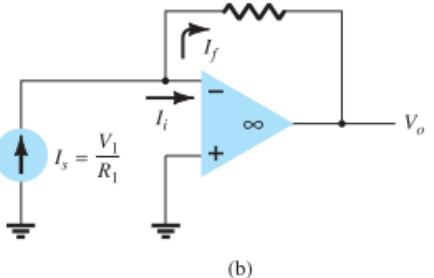
$$A_{vf} = \frac{V_o}{V_s} = \frac{I_o R_C}{V_s} = \left(\frac{I_o}{V_s}\right) R_C = A_f R_C \cong \frac{-h_{fe} R_C}{h_{ie} + h_{fe} R_E}$$

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# Voltage Shunt Feedback



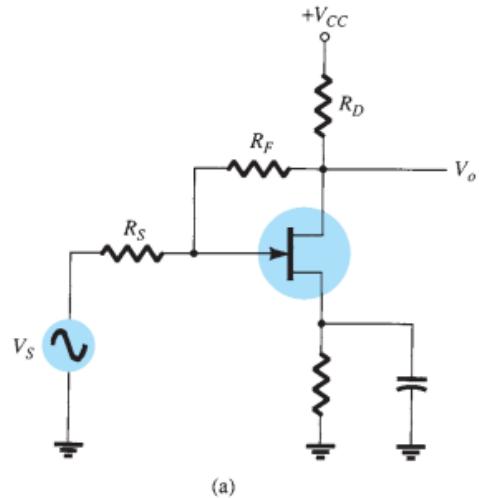
(a)



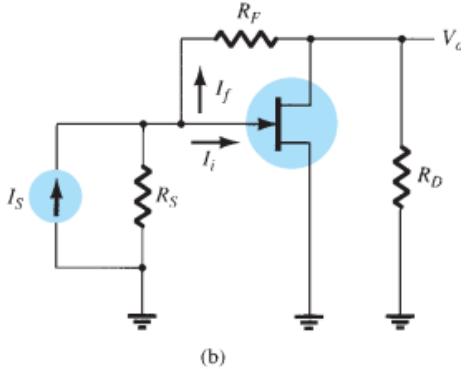
(b)

FIG. 14.12

Voltage-shunt negative feedback amplifier: (a) constant-gain circuit; (b) equivalent circuit.



(a)



(b)

FIG. 14.13

Voltage-shunt feedback amplifier using an FET: (a) circuit; (b) equivalent circuit.

$$A = \frac{V_o}{I_i} = \infty$$

$$\beta = \frac{I_f}{V_o} = \frac{-1}{R_o}$$

$$A_f = \frac{V_o}{I_s} = \frac{V_o}{I_i} = \frac{A}{1 + \beta A} = \frac{1}{\beta} = -R_o$$

$$A_{vf} = \frac{V_o}{I_s} \frac{I_s}{V_1} = (-R_o) \frac{1}{R_1} = \frac{-R_o}{R_1}$$

$$A = \frac{V_o}{I_i} \cong -g_m R_D R_S$$

$$\beta = \frac{I_f}{V_o} = \frac{-1}{R_F}$$

$$A_f = \frac{V_o}{I_s} = \frac{A}{1 + \beta A} = \frac{-g_m R_D R_S}{1 + (-1/R_F)(-g_m R_D R_S)}$$

$$= \frac{-g_m R_D R_S R_F}{R_F + g_m R_D R_S}$$

$$A_{vf} = \frac{V_o}{I_s} \frac{I_s}{V_s} = \frac{-g_m R_D R_S R_F}{R_F + g_m R_D R_S} \left( \frac{1}{R_S} \right)$$

$$= \frac{-g_m R_D R_F}{R_F + g_m R_D R_S} = (-g_m R_D) \frac{R_F}{R_F + g_m R_D R_S}$$

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- For more details, refer to:
  - Chapter 14, Electronic Devices and Circuits, Boylestad.
- The lecture is available online at:
  - [https://speakerdeck.com/ahmad\\_elbanna](https://speakerdeck.com/ahmad_elbanna)
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